



## Application Note

AN000646

# Low Power Proximity Detection

**Minimizing Power Consumption with the TMD2635,  
TMD2636 and TMD2637**

v3-00 • 2021-Mar-26

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# Content Guide

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# 1 Introduction

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## Information

This document applies to the **ams** TMD2635, TMD2636 and TMD2637 devices. References to these devices will be abbreviated to “TMD2635/6/7” throughout.

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The **ams** TMD2635/6/7 was developed for proximity sensing applications requiring small size, low power and short range operation. Example applications include battery powered wearable products such as true wireless stereo earbuds, glasses, watches, etc.

Low power system designers need to strike a balance between competing requirements for optimal performance and minimal power consumption. The TMD2635/6/7 offers features that support low power operation but the application’s required proximity response will ultimately limit the degree to which these features can be leveraged. This application note will describe the settings of the TMD2635/6/7 that affect power consumption and demonstrate how to calculate power consumption based on the sensor’s configuration.



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## Information

For guidance on optimizing the sensor’s proximity response, refer to **ams** application note AN000556 “Proximity Detection – Optimizing Proximity Parameters.”

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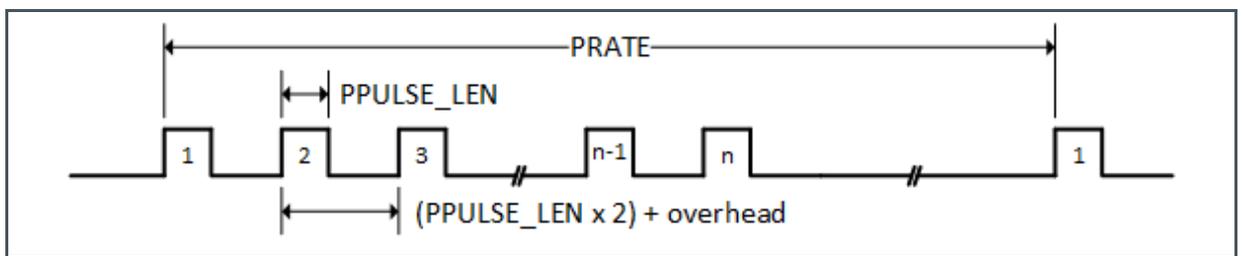
Configuring the TMD2635/6/7 for low power operation requires an understanding of current consumed during all operating states and how to control the duration of those states. Knowing these two things, the “duty cycle” of the sensor can be adjusted. Duty cycle is defined as the fraction of one operating period in which the sensor is active; therefore a “low” duty cycle means that the sensor spends most of its time in a low power state. Ideally, the sensor should operate at full power only as long as is necessary to meet the application’s proximity response requirements.

Duty cycle control of the TMD2635/6/7 can be achieved in a couple of ways. The first and most basic is through the configuration of the sensor’s proximity cycle timing.

## 2 Proximity Cycle Timing & States

When the TMD2635/6/7 is fully operating, it loops in a ‘proximity cycle’ whose timing is controlled by configuration register settings. A single proximity sample (Figure 1) consists of multiple user-programmable pulses (quantity “n”) which enable the IR light source (emitter) when high. In the case of the TMD2635/6/7, the IR emitter is called a “vertical cavity surface emitting laser” (VCSEL) and the IR detector is a photodiode. **PRATE** is the duration of the entire proximity sample.

**Figure 1:**  
Pulse Timing for a Single Proximity Sample



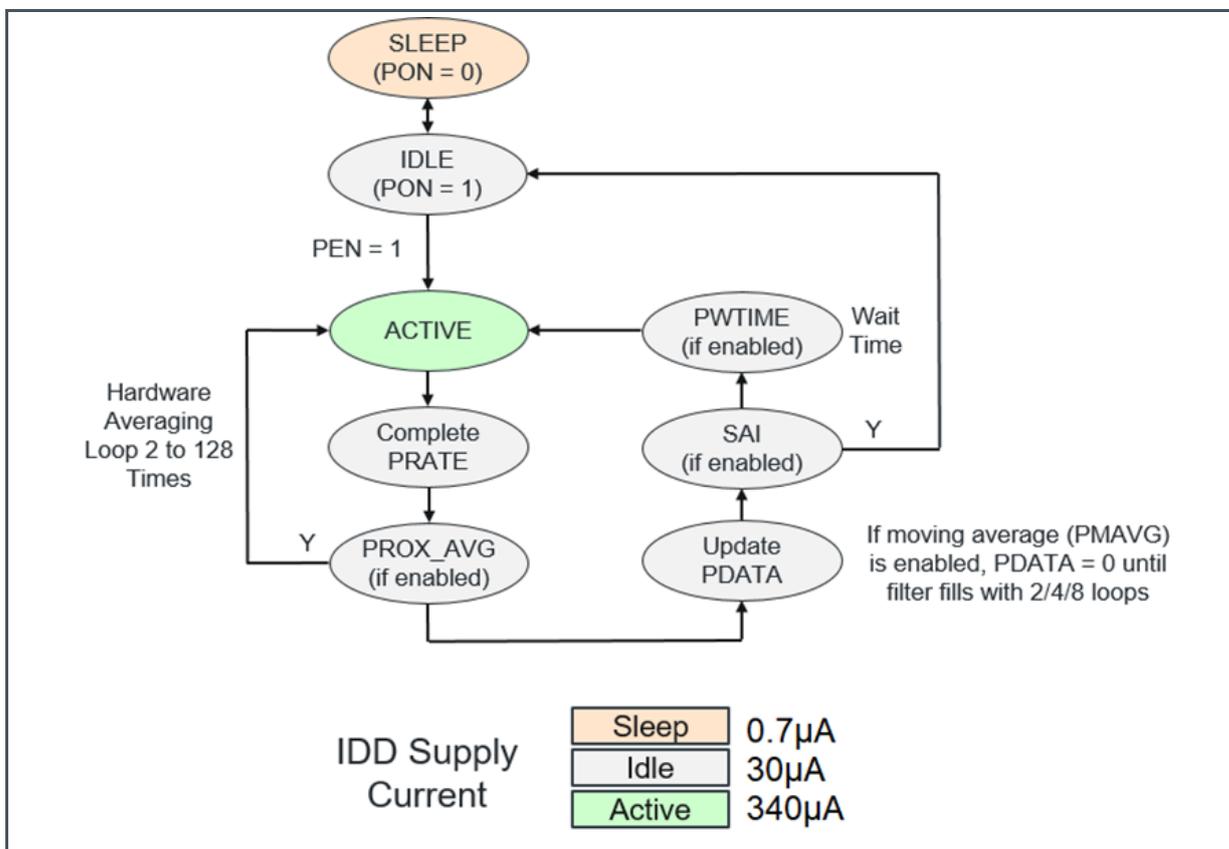
The proximity cycle state diagram (Figure 2) shows the sequence of operation and denotes the average amount of  $I_{DD}$  current consumed during each state. The proximity sample pulsing shown above occurs entirely during the ACTIVE state. This state consumes the most power by far, drawing over 11x more average  $I_{DD}$  current than IDLE states and 485x more than the SLEEP state.

$I_{DD}$  current is sourced from a 1.8 V supply and is consumed by the TMD2635/6/7’s digital circuitry. During the ACTIVE state only, there is another current that is consumed by the TMD2635/6/7’s VCSEL IR emitter. This current is sourced from a VCSEL-dedicated 3.3 V supply and flows only during the high time of each drive pulse. The average value of this current is in the 3.5–15 mA<sup>1</sup> range. Although the drive pulses are narrow, the VCSEL becomes a more significant power consumer as the number and the width of these pulses increase. It is important that the 3.3 V power supply is capable of supporting pulses up to 15 mA in amplitude and 32  $\mu$ s in duration.

$I_{DD}$  and VCSEL current combine during the ACTIVE state to create a major portion of the sensor’s overall power load to the system. Pulse length and the number of pulses should be minimized when configuring the sensor for the required proximity response.

<sup>1</sup> VCSEL drive current is factory-trimmed to achieve a specific proximity response at a given PLDRIVE setting. The actual drive current can be as much as  $\pm 50\%$  of the value listed in the PLDRIVE bit field. This means the nominal 7–10 mA range can actually be 3.5–15 mA.

Figure 2:  
Proximity Cycle Simplified State Diagram



## 3 Configuration Register Settings

There is a specific set of register settings that directly affect both the timing and the power consumption of the proximity cycle.

The following bit fields configure the pulse train of a single proximity sample:

- **PPULSE** (number of pulses, 1–64). Power consumption is directly proportional to the number of pulses.
- **PPULSE\_LEN** (pulse length, 1–32  $\mu$ s). Power consumption is directly proportional to pulse length.
- **PRATE** (proximity sample duration, 88–22528  $\mu$ s). Longer **PRATE** can reduce power consumption by decreasing the duty cycle of the proximity loop.

The following bit fields add optional wait time between proximity samples:

- **PWTIME** (proximity wait time, 2.78–712 ms). Longer wait time can reduce power consumption by decreasing the duty cycle of the proximity loop.
- **PWLONG** (wait time 12 $\times$  extension, 33.36–8544 ms). Longer wait time can reduce power consumption by decreasing the duty cycle of the proximity loop.

The following bit field averages multiple proximity samples into a single result:

- **PROX\_AVG** (hardware averaging, 2–128 samples). Power consumption is directly proportional to the number of hardware averaging repetitions.

Note that the **PMAVG** bit field (proximity moving average, 2, 4 or 8 values) adds delay only at loop start-up and thereafter it has no effect on duty cycle or power consumption.

Although the following bit field does not affect loop timing, it does affect power consumption:

- **PLDRIVE** (VCSEL drive current, 7–10 mA). Power consumption is directly proportional to VCSEL drive strength

Once the optimal proximity response has been established, potential power savings can be explored by experimenting with reductions in duty cycle and drive current.

## 4 Application Software Control

A second level of duty cycle control exists at the application software level. The host processor can be programmed to cycle the sensor's active/inactive states in either a "polled" or an "interrupt-driven" fashion.

A polling cycle consists of waking up the sensor, sampling data then putting the sensor back to sleep.

An interrupt-driven cycle consists of waking up the sensor, reading its previous sample then letting it free-run. When the next data event occurs, the sensor signals an interrupt to the host and automatically goes into a low power state.

The interrupt cycle described above employs the TMD2635/6/7 "Sleep after Interrupt" (**SAI**) feature. Referring to Figure 2, when the IDLE state is entered following an **SAI**-enabled interrupt, the oscillator is deactivated to reduce power consumption. The device can only be reactivated by clearing interrupt flags in the **STATUS** register.

TMD2635/6/7 interrupt events are programmable. Refer to the latest datasheets for a description of the following features that configure interrupt trigger conditions:

- Proximity High and Low Thresholds – Generate an interrupt when **PDATA** falls outside of the threshold range and the persistence filter value is reached. See the **PILT**, **PILTH**, **PIHTL** and **PIHTH** register descriptions in the TMD2635/6/7 datasheets.
- Persistence Filter – Counts consecutive occurrences of **PDATA** values that fall outside of the threshold range. See the **PERS** register description in the TMD2635/6/7 datasheets.
- Proximity Interrupt Mode (PIM) – Qualifies interrupt assertion based on the threshold crossing's level or state change. See the **INTENAB** register description and Figure 43 in the TMD2635/6/7 datasheets.

Timing in the interrupt-driven scenario is less deterministic than in a polled scenario. Event-driven duty cycles depend on host response time and the changing stimulus in the sensor's field of view that trigger interrupts based on the register configuration described above. This timing variability makes accurate power calculation difficult unless simplifying assumptions are made. Bench characterization is the best way to accurately assess power consumption under such dynamic conditions.

Assuming that the host will service interrupts promptly, the sensor will spend a majority of its time in states that draw idle current (30  $\mu$ A average) while free-running, waiting for the next interrupt trigger event to drive it into a low power state. This interrupt-driven approach will *not* be able to conserve power as well as a polled design that purposely maximizes the time spent in SLEEP state drawing 0.7  $\mu$ A average current.

## 5 Power Calculation Examples

### 5.1 Proximity Cycle Power Calculations

The TMD2635/6/7 datasheets provide all of the information needed to calculate average current consumed during the sensor's proximity cycle. For example, consider a sensor configured as follows:

**Figure 3:**  
**Device Configuration Proximity Cycle**

Device Configuration	Inputs	Units	Range, Notes
VDD3 Voltage	3.0	V	2.9 V to 3.6 V
VDD Voltage	1.8	V	1.7 V to 2.0 V
IDC Current SLEEP state	0.7	μA	SLEEP state occurs when <b>PON</b> = 0 and I <sup>2</sup> C bus is idle. If SLEEP state has been entered as the result of operational flow, <b>SAI</b> = 1 and <b>PON</b> will remain high.
IDC Current IDLE state	30	μA	IDLE state occurs when <b>PON</b> = 1 and the device is not in the ACTIVE state.
IDC Current ACTIVE state	340	μA	ACTIVE state occurs when <b>PON</b> = 1 and the device is actively integrating.
VCSEL Current ( <b>PLDRIVE</b> )	7	mA	7 mA to 10 mA
Number of Pulses ( <b>PPULSE</b> )	4	x	1 to 64
Pulse Length ( <b>PPULSE_LEN</b> )	16	μs	1/2/4/8/12/16/24/32 μs
Wait Time ( <b>PWTIME</b> )	2780	μs	2.78 ms to 712 ms or 33.36 ms to 8544 ms when <b>PWLONG</b> =1, $((\mathbf{PWTIME}=0..255)+1)*2.78$ ms
<b>PRATE</b>	2816	μs	88 μs to 22528 μs, $((\mathbf{PRATE}=0..255)+1)*88$ μs, If <b>PRATE</b> is programmed to a value that is less than the time required for the selected number of pulses, then it is ignored and the proximity cycle duration is extended to allow for the full number of pulses programmed.
Hardware Averaging ( <b>PROX_AVG</b> )	1	x	1/2/4/8/16/32/64/128 (set to 1 if disabled)

Based on the sensor configuration above, duty cycle time calculations can be made.

**Figure 4:**  
**Loop Calculations Proximity Cycle**

Per Loop Time Calculations	Outputs	Units	Formula
Active Time per Proximity Cycle	406.75	µs	$= (7 \times \text{PPULSE\_LEN}) + (\text{PPULSE} \times ((2 \times \text{PPULSE\_LEN}) + 22 \mu\text{s})) + 78.75 \mu\text{s}$
Total Active Time per Loop	406.75	µs	$= (\text{Active Time per Proximity Cycle}) \times (\text{PROX\_AVG})$
Idle Time (all time except active)	5189.25	µs	$= (\text{PWTIME}) + [\text{if } \text{PRATE} > (\text{Active Time per Proximity Cycle}), \text{ then } ((\text{PRATE} - (\text{Active Time per Proximity Cycle})) \times (\text{PROX\_AVG})), \text{ else } 0]$
Total Loop Time	5596	µs	$= (\text{Total Active Time per Loop}) + (\text{Idle Time})$

Applying the duty cycle time calculations above to the appropriate power supplies will determine the average current drawn from each source.

**Figure 5:**  
**Power Calculations Proximity Cycle**

Power Calculations	Outputs	Units	Formula
Average VCSEL Current	80.06	µA	$= (\text{PLDRIVE} \times 1000) \times ((\text{PPULSE} \times \text{PPULSE\_LEN} \times \text{PROX\_AVG}) / (\text{Total Loop Time}))$
Average VCSEL Power	240.17	µW	$= (\text{Average VCSEL Current}) \times (\text{VDD3 Voltage})$
Average IDD Current	52.53	µA	$= [(\text{IDD Current ACTIVE state}) \times ((\text{Total Active Time per Loop}) / (\text{Total Loop Time}))] + [(\text{IDD Current IDLE state}) \times ((\text{Idle Time}) / (\text{Total Loop Time}))]$
Average IDD Power	94.56	µW	$= (\text{Average IDD Current}) \times (\text{VDD Voltage})$
Total Average Power Consumption	334.73	µW	$= (\text{Average IDD Power}) + (\text{Average VCSEL Power})$

The calculations demonstrated above are made based on the proximity cycle of the sensor while it is fully operational and free-running.

## 5.2 Polling Cycle Power Calculations

The following example calculates the average power consumed by a hypothetical polling cycle that consists of waking up the sensor, collecting a single data sample from 1 proximity cycle then putting the sensor back to sleep. Consider the following sensor configuration with software-controlled duty cycle timing:

**Figure 6:**  
**Device Configuration Polling Cycle**

Device Configuration	Inputs	Units	Range, Notes
VDD3 Voltage	3	V	2.9 V to 3.6 V
VDD Voltage	1.8	V	1.7 V to 2.0 V
IDD Current - SLEEP state	0.7	μA	SLEEP state occurs when <b>PON</b> = 0 and I <sup>2</sup> C bus is idle. If SLEEP state has been entered as the result of operational flow, SAI = 1 and <b>PON</b> will remain high.
IDD Current - IDLE state	30	μA	IDLE state occurs when <b>PON</b> = 1 and the device is not in the ACTIVE state.
IDD Current - ACTIVE state	340	μA	ACTIVE state occurs when <b>PON</b> = 1 and the device is actively integrating.
VCSEL Current	7	mA	7 mA to 10 mA
Number of Pulses ( <b>PPULSE</b> )	4	x	1 to 64
Pulse Length ( <b>PPULSE_LEN</b> )	16	μs	1/2/4/8/12/16/24/32 μs
Time Between Proximity Samples	250	ms	Controlled by software routine
Wakeup Time	100	μs	
Time Post Active Until Sleep	500	μs	Controlled by software routine

Based on the sensor configuration above, duty cycle time calculations can be made.

**Figure 7:**  
**Time Calculations Polling Cycle**

Per Loop Time Calculations	Outputs	Units	Formula
Active Time	406.75	μs	$= (7 \times \text{PPULSE\_LEN}) + \text{PPULSE} \times (2 \times \text{PPULSE\_LEN} + 22 \mu\text{s}) + 78.75 \mu\text{s}$
Idle Time (wakeup + post active time)	600	μs	$= (\text{Wakeup Time}) + (\text{Time Post Active Until Sleep})$

Per Loop Time Calculations	Outputs	Units	Formula
Sleep Time	248993.3	μs	= (Time Between Proximity Samples x 1000) - (Active Time) - (Idle Time)
Total Loop Time	250000	μs	= (Active Time) + (Idle Time) + (Sleep Time)

In the polling cycle described above, the host interaction with the sensor generates bus activity that consumes system current through the I<sup>2</sup>C bus pull-up resistors. This power can be accounted for in the sensor’s total power calculation.

To analyze I<sup>2</sup>C bus activity, consider the following assumptions about the steps taken by the polling software routine, including the number of 9-bit bytes transferred during each transaction<sup>2</sup>

1. Write the **ENABLE** register to wake up the sensor (3 bytes)
2. Wait at least (Wakeup Time + Active Time) for a proximity cycle to occur
3. Read the **PDATA1** and **PDATAH** registers (5 bytes)
4. Write the **ENABLE** register to put sensor to sleep (3 bytes)

This short routine generates 11 bytes of bus activity.

Consider the following I<sup>2</sup>C bus parameters:

**Figure 8:**  
I<sup>2</sup>C Bus Parameters

I <sup>2</sup> C Bus Parameters	Inputs	Units	Notes
I <sup>2</sup> C Bus Voltage	1.8	V	
Pull Up Resistance	2200	Ohms	Value based on the I <sup>2</sup> C bus voltage, system bus speed and trace capacitance
I <sup>2</sup> C Bus Speed	200000	Hz	Assume 50% SCL duty cycle
Number of Bytes per Cycle	11		

Applying the duty cycle time calculations above to the appropriate power supplies will determine the average current drawn from each source. Using the I<sup>2</sup>C bus parameters above, bus current and power consumed during each polling cycle can be estimated.

<sup>2</sup> More details on I<sup>2</sup>C read and write transactions can be found in the TMD2635/6/7 datasheets.

**Figure 9:**  
**Power Calculations Polling Cycles**

Power Calculations	Outputs	Units	Formula
Average VCSEL Current	1.79	μA	= (VCSEL Current x 1000) x (PPULSE x PPULSE_LEN) / (Total Loop Time)
Average VCSEL Power	5.38	μW	= (Average VCSEL Current) x (VDD3 Voltage)
Average IDD Current	1.32	μA	= (IDD Current ACTIVE State) x (Active Time / Total Loop Time) + (IDD Current SLEEP State) x (Sleep Time / Total Loop Time) + (IDD Current IDLE State) x (Idle Time / Total Loop Time)
Average IDD Power	2.38	μW	= (Average IDD Current) x (VDD Voltage)
I <sup>2</sup> C Bus Current per Polling Cycle <sup>(1)</sup>	0.41	μA	= [((I <sup>2</sup> C Bus Voltage) / (Pull up resistance)) x 1e6] x [((Number of Bytes per Cycle) x (9 bits per Byte)) x (1 / (I <sup>2</sup> C Bus Speed))]
I <sup>2</sup> C Bus Power per Polling Cycle	0.73	μW	= (Bus Current per Polling Cycle) x (I <sup>2</sup> C Bus Voltage)
Total Power Consumption	8.49	μW	= (Average VCSEL Power) + (Average IDD Power) + (I <sup>2</sup> C Bus Power)

<sup>(1)</sup> Using the full bit period in the current calculation accounts for both the SCL and SDA lines assuming that they both maintain a 50% duty cycle and that half of the data bits are low during each bus transaction. Actual bus power will vary with data value

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## 6 Summary

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Implementing proximity detection in a low power system should begin by optimizing the sensor response according to the system detect/release requirements. Power savings can then be explored by experimenting with changes in the proximity cycle timing, emitter drive current and duty cycling at the application software level. A polled sensor design will provide more predictable duty cycle control that can achieve lower average power consumption than an interrupt-driven approach. Calculation of TMD2635/6/7 average power consumption can be made using the techniques outlined in this document.

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## 7 References

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**For further information, please refer to the following documents:**

- TMD2635 Datasheet
  - TMD2636 Datasheet
  - TMD2637 Datasheet
  - AN000556 Proximity Detection – Optimizing Proximity Parameters
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## 8 Revision Information

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Changes from previous version to current revision v3-00	Page
Changed TMD2635/6 to TMD2635/6/7	3,4,7,8,11,13
Added TMD2637 datasheet to References	14

- Page and figure numbers for the previous version may differ from page and figure numbers in the current revision.
- Correction of typographical errors is not explicitly mentioned.

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